

第一次作业

1.5(2)

判断下列各序列是否为周期性的. 若是, 确定其周期

$$f_2(k) = \cos\left(\frac{3\pi}{4}k + \frac{\pi}{4}\right) + \cos\left(\frac{\pi}{3}k + \frac{\pi}{6}\right)$$

解:  $\cos\left(\frac{3\pi}{4}k + \frac{\pi}{4}\right)$  的周期为  $2\pi \cdot \frac{4}{3\pi} = \frac{T}{N}$ ,  $N$  为一整数,  $T$  为满足等式关系的最小正整数, 故  $T_1 = 8$ .

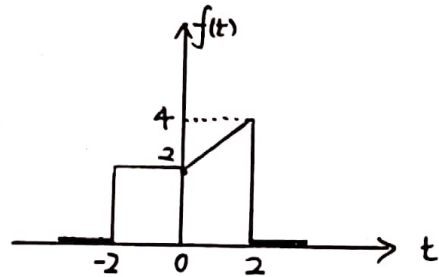
同理, 对  $\cos\left(\frac{\pi}{3}k + \frac{\pi}{6}\right)$  的周期  $2\pi \cdot \frac{3}{\pi} = \frac{T}{N}$ ,  $T_2 = 6$ .

$f_2(k)$  周期应为  $T_1, T_2$  的最小公倍数即  $T = 24$ .

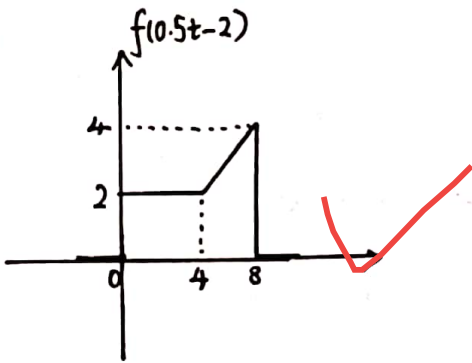
1.6(b)(8) 已知信号波形如下图, 画出下列各波形函数

(b)  $f(0.5t-2)$

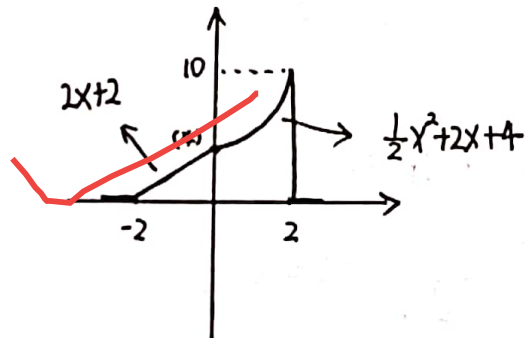
(8)  $\int_{-\infty}^t f(x) dx$



解: (b)



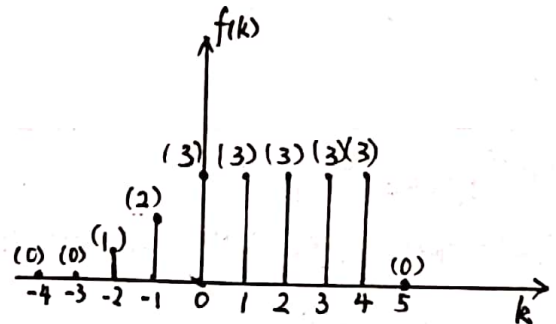
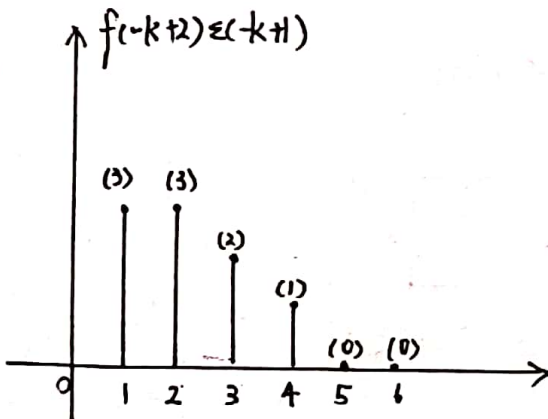
(8)



1.7 已知序列  $f(k)$  图形如下, 画出下列序列图形

(5)  $f(-k+2) \varepsilon(-k+1)$

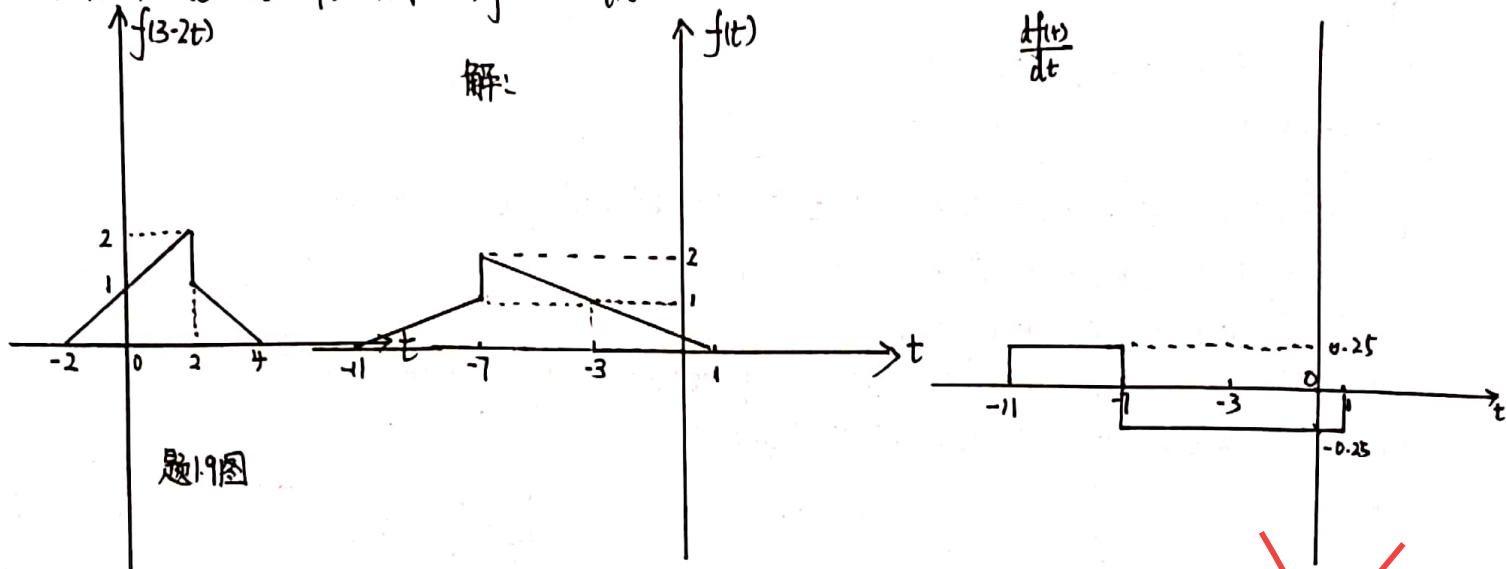
解:



题图



1.9. 已知信号波形如下, 分别画出  $f(t)$  和  $\frac{df(t)}{dt}$  的波形.



1.10 (3)(7) 计算

(3)  $\int_{-\infty}^{\infty} \frac{\sin(\pi t)}{t} \delta(t) dt$     (7)  $\int_{-\infty}^{\infty} (t^3 + 2t^2 - 2t + 1) \delta'(t-1) dt$

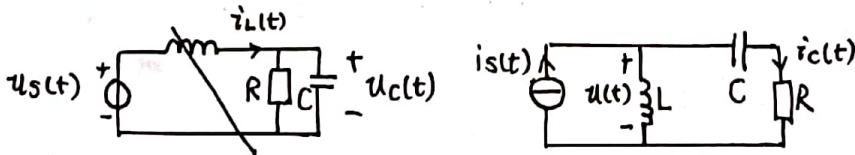
解:

(3)  $\int_{-\infty}^{\infty} \frac{\sin(\pi t)}{t} \delta(t) dt$   
 $= \lim_{t \rightarrow 0} \frac{\sin(\pi t)}{t} = \lim_{t \rightarrow 0} \frac{\pi \cos(\pi t)}{1} = \pi$

(7)  $\int_{-\infty}^{\infty} (t^3 + 2t^2 - 2t + 1) \delta'(t-1) dt$   
 $= -(t^3 + 2t^2 - 2t + 1)' \Big|_{t=1}$   
 $= -(3t^2 + 4t - 2) \Big|_{t=1}$   
 $= -5$

如图 1.13 所示电路, 写出

1.13 (1) 以  $u_c(t)$  为响应的微分方程 (2) 以  $i_c(t)$  为响应的微分方程



解: (1) 设电容 C 电压为  $u_c(t)$ .  
 电路电流  $i_L(t)$ .  
 $u(t) = u_c(t) + i_c(t)R$   
 $u'(t) = u_c'(t) + i_c'(t)R$   
 $i_c(t) = C u_c'(t)$

$\therefore u_c'(t) = \frac{i_c(t)}{C}$  ①  
 $\therefore i_c(t) = i_s(t) - i_L(t)$   
 $u(t) = L i_L'(t)$   
 $\therefore i_c'(t) = i_s'(t) - i_L'(t)$   
 $= i_s'(t) - \frac{u(t)}{L}$  ②

(2) ② 式代入  $u'(t)$  表达式

$u'(t) = \frac{i_c(t)}{C} + i_s'(t)R - \frac{u(t)}{L}R$

两端求导

得  $u''(t) = \frac{1}{C} i_c'(t) + i_s''(t)R - \frac{u'(t)}{L}R$

即  $u''(t) = \frac{1}{C} i_s'(t) - \frac{u(t)}{CL} + i_s''(t)R - \frac{u'(t)}{L}R$



扫描全能王 创建

∴ 以  $u(t)$  为响应的微分方程为

$$u''(t) + \frac{R}{L} u'(t) + \frac{1}{LC} u(t) = \frac{1}{C} i_s'(t) + R i_s''(t)$$

(2) 以  $i_c(t)$  为响应的微分方程

$$i_c(t) = i_s(t) - i_L(t)$$

$$i_c'(t) = i_s'(t) - i_L'(t)$$

$$\because i_L'(t) = \frac{u(t)}{L},$$

$$\therefore i_c'(t) = i_s'(t) - \frac{u(t)}{L}$$

$$\text{对两端求导有 } i_c''(t) = i_s''(t) - \frac{u'(t)}{L}$$

$$\because u'(t) = u_c'(t) + i_c'(t)R$$

$$\text{A } u_c'(t) = \frac{i_c(t)}{C}, \quad \therefore u'(t) = \frac{i_c(t)}{C} + i_c'(t)R$$

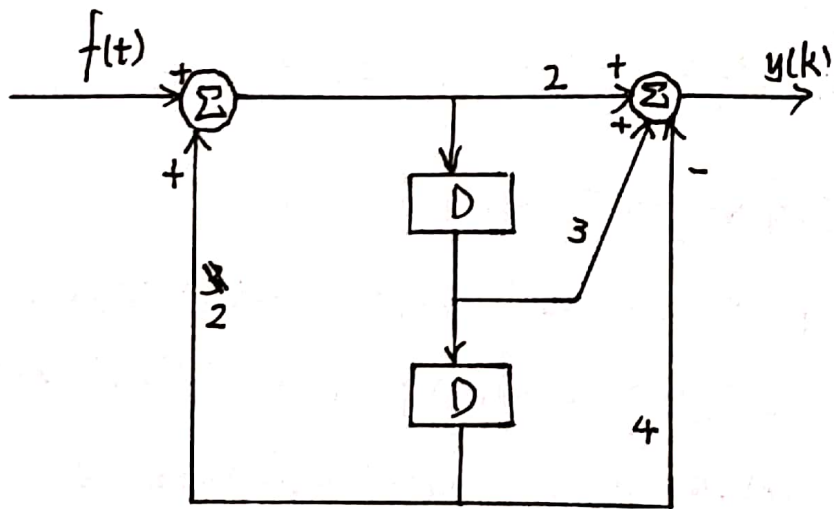
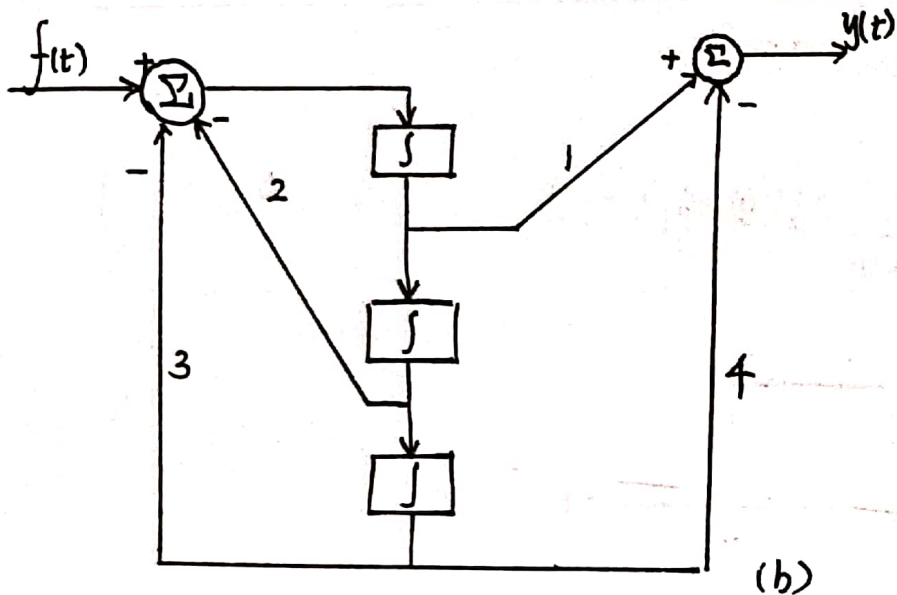
$$\therefore i_c''(t) = i_s''(t) - \frac{1}{LC} i_c(t) - \frac{R}{L} i_c'(t)$$

故以  $i_c(t)$  为响应的微分方程为

$$i_c''(t) + \frac{R}{L} i_c'(t) + \frac{1}{LC} i_c(t) = i_s''(t)$$



1.20 (b)(d) 写出下面图中各系统的微分方程和差分方程



解:

(b) 设最下方积分器输出为  $x(t)$ .

左端加法器输出  $x^{(3)}(t) = f(t) - 2x^{(1)}(t) - 3x(t)$

$$f(t) = x^{(3)}(t) + 2x^{(1)}(t) + 3x(t)$$

右端加法器输出  $y(t) = x^{(2)}(t) - 4x(t)$

$$\begin{cases} y^{(3)}(t) + 2y^{(1)}(t) + 3y(t) = [x^{(3)}(t) + 2x^{(1)}(t) + 3x(t)]'' - 4[x^{(3)}(t) + 2x^{(1)}(t) + 3x(t)] \\ 2y^{(1)}(t) = [2x^{(1)}(t)]'' - 4[2x^{(1)}(t)] \\ 3y(t) = [3x(t)]'' - 4[3x(t)] \end{cases}$$

$$y^{(3)}(t) + 2y^{(1)}(t) + 3y(t) = [x^{(3)}(t) + 2x^{(1)}(t) + 3x(t)]'' - 4[x^{(3)}(t) + 2x^{(1)}(t) + 3x(t)]$$

$$\text{即 } y^{(3)}(t) + 2y^{(1)}(t) + 3y(t) = f^{(2)}(t) - 4f(t)$$

此为系统微分方程.



(d). 设最上方延迟单元输入为  $x(k)$ .

左方加法器输出  $x(k) = f(k) + 2x(k-2)$

即  $f(k) = x(k) - 2x(k-2)$ .

右方加法器输出

$$y(k) = 2x(k) + 3x(k-1) - 4x(k-2)$$

$$-2y(k-2) = 2[-2x(k-2)] + 3[-2x(k-3)] - 4[-2x(k-4)]$$

相加得

$$y(k) - 2y(k-2) = 2[x(k) - x(k-2)] + 3[x(k-1) - 2x(k-3)] - 4[x(k-2) - 2x(k-4)]$$

即

$$y(k) - 2y(k-2) = 2f(k) + 3f(k-1) - 4f(k-2)$$



1.27 某LTI连续系统,其初始状态一定,已知当激励为 $f(t)$ 时,其全响应为

$$y_1(t) = e^{-t} + \cos(\pi t), t \geq 0$$

若初始状态不变,激励为 $2f(t)$ 时,其全响应为

$$y_2(t) = 2\cos(\pi t), t \geq 0$$

求初始状态不变而激励为 $3f(t)$ 时系统的全响应。

解:令系统零状态响应为 $y_{zs}(t)$ ,零输入响应为 $y_{zi}(t)$   
激励为 $f(t)$ 时,

$$y_{zi}(t) + y_{zs}(t) = e^{-t} + \cos(\pi t), t \geq 0$$

∵ LTI系统具有齐次性

$$2y_{zs}(t) = T[2f(t)], 3y_{zs}(t) = T[3f(t)]$$

初始状态不变,激励为 $2f(t)$ 时,全响应

$$y_{zi}(t) + 2y_{zs}(t) = 2\cos(\pi t), t \geq 0$$

解得

$$y_{zi}(t) = 2e^{-t}$$

$$y_{zs}(t) = -e^{-t} + \cos(\pi t), t \geq 0$$

初始状态不变,激励为 $3f(t)$ 时系统的全响应为

$$y_{zi}(t) + y_{zs}(t) = -e^{-t} + 3\cos(\pi t), t \geq 0$$

即激励为 $3f(t)$ 时全响应

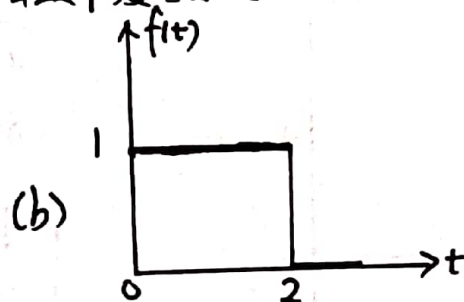
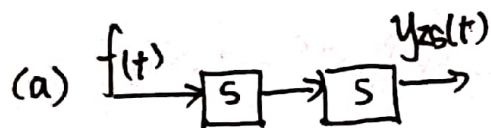
$$y_3(t) = -e^{-t} + 3\cos(\pi t), t \geq 0$$



1.31 如有LTI连续系统S, 已知当激励为阶跃函数  $\varepsilon(t)$  时, 其零状态响应为

$$\varepsilon(t) - 2\varepsilon(t-1) + \varepsilon(t-2)$$

现将两个完全相同的系统相级联, 如1.31图(a)示. 当这个复合系统的输入为题1.31(b)所示信号  $f(t)$  时, 求该系统的零状态响应.



解: 对系统S输入信号  $\varepsilon(t)$ , 输出响应

$$y_{fs}(t) = \varepsilon(t) - 2\varepsilon(t-1) + \varepsilon(t-2)$$

令  $\varepsilon(t)$  作为输入信号, 得系统S输出为

$$y_{fs}^{(2)}(t) = y_{fs}(t) - 2y_{fs}(t-1) + y_{fs}(t-2)$$

$$= \varepsilon(t) - 4\varepsilon(t-1) + 6\varepsilon(t-2) - 4\varepsilon(t-3) + \varepsilon(t-4)$$

当复合系统输入信号  $\varepsilon(t)$  时, 其输出响应

$$y(t) = \varepsilon(t) - 4\varepsilon(t-1) + 6\varepsilon(t-2) - 4\varepsilon(t-3) + \varepsilon(t-4)$$

1.31(b)的输入信号为  $\varepsilon(t) - \varepsilon(t-2)$

∵ LTI系统具有齐次性、可加性与时不变性,

得复合系统的零状态响应

$$\begin{aligned} y_f(t) &= [\varepsilon(t) - 4\varepsilon(t-1) + 6\varepsilon(t-2) - 4\varepsilon(t-3) + \varepsilon(t-4)] \\ &\quad - [\varepsilon(t-2) - 4\varepsilon(t-3) + 6\varepsilon(t-4) - 4\varepsilon(t-5) + \varepsilon(t-6)] \\ &= \varepsilon(t) - 4\varepsilon(t-1) + 5\varepsilon(t-2) - 5\varepsilon(t-4) + 4\varepsilon(t-5) - \varepsilon(t-6) \end{aligned}$$

